

## FEATURES

- Single-supply operation: 3.0 V to 30 V
- Wide input voltage range
- Rail-to-rail output swing
- Low supply current: 200  $\mu$ A/amplifier
- Wide bandwidth: 1.2 MHz
- Slew rate: 0.46 V/ $\mu$ s
- Low offset voltage: 250  $\mu$ V maximum
- No phase reversal
- Overshoot protection (OVP)
  - 25 V above/below supply rails at  $\pm$ 5 V
  - 12 V above/below supply rails at  $\pm$ 15 V

## APPLICATIONS

- Industrial process control
- Battery-powered instrumentation
- Power supply control and protection
- Telecommunications
- Remote sensors
- Low voltage strain gage amplifiers
- DAC output amplifiers

## GENERAL DESCRIPTION

The ADA4091-2 dual and ADA4091-4 quad are micropower, single-supply, 1.2 MHz bandwidth amplifiers featuring rail-to-rail inputs and outputs. They are guaranteed to operate from a +3.0 V to +30 V single supply as well as from  $\pm$ 1.5 V to  $\pm$ 15 V dual supplies.

The ADA4091-2/ADA4091-4 features a unique input stage that allows the input voltage to exceed either supply safely without any phase reversal or latch-up; this is called overshoot protection (OVP).

Applications for these amplifiers include portable telecommunications equipment, power supply control and protection, and interface for transducers with wide output ranges. Sensors requiring a rail-to-rail input amplifier include Hall effect, piezo-electric, and resistive transducers.

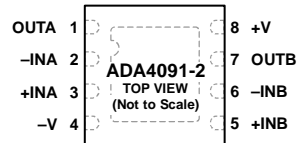
The ability to swing rail-to-rail at both the input and output enables designers, for example, to build multistage filters in single-supply systems and to maintain high signal-to-noise ratios (SNR).

The ADA4091-2/ADA4091-4 is specified over the extended industrial temperature range of  $-40^{\circ}\text{C}$  to  $+125^{\circ}\text{C}$ . The ADA4091-2/ADA4091-4 is part of the growing selection of 36 V, low power operational amplifiers from Analog Devices, Inc., (see Table 1).

## PIN CONFIGURATIONS



Figure 1. 8-Lead, Narrow-Body SOIC (R-8)



- NOTES
1. IT IS RECOMMENDED TO CONNECT THE EXPOSED PAD TO V-.

Figure 2. 8-Lead LFCSP (CP-8-21)

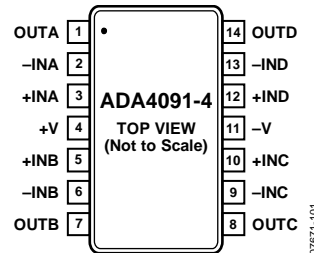
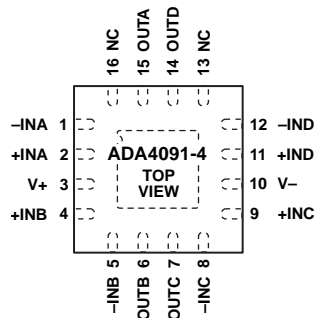


Figure 3. 14-Lead TSSOP (RU-14)



- NOTES
1. NC = NO CONNECT.
  2. IT IS RECOMMENDED TO CONNECT THE EXPOSED PAD TO V-.

Figure 4. 16-Lead LFCSP (CP-16-17)

The ADA4091-2 is available in 8-lead, plastic SOIC and 8-lead LFCSP packages. The ADA4091-4 is available in 14-lead TSSOP and 16-lead LFCSP surface-mount packages.

Table 1. Low Power, 36 V Operational Amplifiers

Family	Rail-to-Rail I/O	PJFET	Low Noise
Single			OP1177
Dual	ADA4091-2	AD8682	OP2177
Quad	ADA4091-4	AD8684	OP4177

## Rev. H

## Document Feedback

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**REVISION HISTORY**

**5/2016—Rev. G. to Rev. H**

Changed CP-8-9 to CP-8-21 ..... Throughout  
 Changes to Figure 2 ..... 1  
 Updated Outline Dimensions ..... 16  
 Changes to Ordering Guide ..... 18

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 Updated Outline Dimensions ..... 16  
 Changes to Ordering Guide ..... 18

**7/2009—Rev. A to Rev. B**

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 Changes to Table 2 ..... 3  
 Changes to Table 3 ..... 4  
 Changes to Table 4 ..... 5  
 Changes to Table 5 ..... 6  
 Changes to Table 6 ..... 6  
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 Changes to Ordering Guide ..... 16

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Changes to Data Sheet Title ..... 1  
 Changes to Features ..... 1  
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 Added New Figure 43 ..... 13  
 Changes to Input Overvoltage Protection Section ..... 15  
 Changes to Ordering Guide ..... 16

**10/2008—Revision 0: Initial Version**

## SPECIFICATIONS

## ELECTRICAL SPECIFICATIONS

$V_{SY} = \pm 1.5$  V,  $V_{CM} = 0.0$  V,  $T_A = 25^\circ\text{C}$ , unless otherwise noted.

Table 2.

Parameter	Symbol	Test Conditions/Comments	Min	Typ	Max	Unit
INPUT CHARACTERISTICS						
Offset Voltage	$V_{OS}$		-250	-40	+250	$\mu\text{V}$
		ADA4091-4 LFCSP package	-400	-40	+400	$\mu\text{V}$
		$-40^\circ\text{C} \leq T_A \leq +125^\circ\text{C}$	-600		+600	$\mu\text{V}$
Offset Voltage Drift	$\Delta V_{OS}/\Delta T$		2.5		$\mu\text{V}/^\circ\text{C}$	
Input Bias Current	$I_B$		-55	-44		nA
		$-40^\circ\text{C} \leq T_A \leq +85^\circ\text{C}$	-55		+55	nA
		$-40^\circ\text{C} \leq T_A \leq +125^\circ\text{C}$	-275		+275	nA
Input Offset Current	$I_{OS}$		-3	0.5	+3	nA
		$-40^\circ\text{C} \leq T_A \leq +85^\circ\text{C}$	-5		+5	nA
		$-40^\circ\text{C} \leq T_A \leq +125^\circ\text{C}$	-75		+75	nA
Input Voltage Range			-1.5		+1.5	V
Common-Mode Rejection Ratio	CMRR	$V_{CM} = -1.35$ V to $+1.35$ V	84	100		dB
		$-40^\circ\text{C} \leq T_A \leq +125^\circ\text{C}$	78			dB
Large Signal Voltage Gain	$A_{VO}$	$R_L = 100$ k $\Omega$ , $V_O = -1.2$ V to $+1.2$ V	106	113		dB
		$-40^\circ\text{C} \leq T_A \leq +125^\circ\text{C}$	101			dB
		$R_L = 10$ k $\Omega$ , $V_O = -1.2$ V to $+1.2$ V	92	94		dB
		$-40^\circ\text{C} \leq T_A \leq +125^\circ\text{C}$	85			dB
OUTPUT CHARACTERISTICS						
Output Voltage High	$V_{OH}$	$R_L = 100$ k $\Omega$ to GND	1.490	1.495		V
		$-40^\circ\text{C} \leq T_A \leq +125^\circ\text{C}$	1.490			V
		$R_L = 10$ k $\Omega$ to GND	1.475	1.485		V
		$-40^\circ\text{C}$ to $+125^\circ\text{C}$	1.455			V
Output Voltage Low	$V_{OL}$	$R_L = 100$ k $\Omega$ to GND		-1.499	-1.495	V
		$-40^\circ\text{C} \leq T_A \leq +125^\circ\text{C}$			-1.495	V
		$R_L = 10$ k $\Omega$ to GND		-1.495	-1.490	V
		$-40^\circ\text{C} \leq T_A \leq +125^\circ\text{C}$			-1.490	V
Short-Circuit Limit	$I_{SC}$	Source/sink		$\pm 31$		mA
Closed-Loop Impedance	$Z_{OUT}$	$f = 1$ MHz, $A_V = 1$		102		$\Omega$
POWER SUPPLY						
Power Supply Rejection Ratio	PSRR	$V_{SY} = 2.7$ V to 36 V	108	126		dB
		$-40^\circ\text{C} \leq T_A \leq +125^\circ\text{C}$	100			dB
Supply Current per Amplifier	$I_{SY}$	$I_O = 0$ mA		165	200	$\mu\text{A}$
		$-40^\circ\text{C} \leq T_A \leq +125^\circ\text{C}$			300	$\mu\text{A}$
DYNAMIC PERFORMANCE						
Slew Rate	SR	$R_L = 100$ k $\Omega$ , $C_L = 30$ pF		0.46		V/ $\mu\text{s}$
Settling Time	$t_s$	To 0.01%		22		$\mu\text{s}$
Gain Bandwidth Product	GBP			1.22		MHz
Phase Margin	$\Phi_M$			69		Degrees
NOISE PERFORMANCE						
Voltage Noise	$e_n$ p-p	0.1 Hz to 10 Hz		0.8		$\mu\text{V}$ p-p
Voltage Noise Density	$e_n$	$f = 1$ kHz		24		nV/ $\sqrt{\text{Hz}}$

$V_{SY} = \pm 5.0\text{ V}$ ,  $V_{CM} = 0.0\text{ V}$ ,  $T_A = 25^\circ\text{C}$ , unless otherwise noted.

Table 3.

Parameter	Symbol	Test Conditions/Comments	Min	Typ	Max	Unit
<b>INPUT CHARACTERISTICS</b>						
Offset Voltage	$V_{OS}$		-250	-45	+250	$\mu\text{V}$
		ADA4091-4 LFCSP package	-400	-40	+400	$\mu\text{V}$
		$-40^\circ\text{C} \leq T_A \leq +125^\circ\text{C}$	-600		+600	$\mu\text{V}$
Offset Voltage Drift	$\Delta V_{OS}/\Delta T$			2.5		$\mu\text{V}/^\circ\text{C}$
Input Bias Current	$I_B$		-60	-50		nA
		$-40^\circ\text{C} \leq T_A \leq +85^\circ\text{C}$	-80		+80	nA
Input Offset Current	$I_{OS}$	$-40^\circ\text{C} \leq T_A \leq +125^\circ\text{C}$	-350		+350	nA
			-3	0.5	+3	nA
Input Voltage Range		$-40^\circ\text{C} \leq T_A \leq +85^\circ\text{C}$	-7		+7	nA
		$-40^\circ\text{C} \leq T_A \leq +125^\circ\text{C}$	-100		+100	nA
Common-Mode Rejection Ratio	CMRR	$V_{CM} = -4.85\text{ V to }+4.85\text{ V}$	95	113		dB
		$-40^\circ\text{C} \leq T_A \leq +125^\circ\text{C}$	88			dB
Large Signal Voltage Gain	$A_{VO}$	$R_L = 100\text{ k}\Omega$ , $V_O = \pm 4.7\text{ V}$	113	117		dB
		$-40^\circ\text{C} \leq T_A \leq +125^\circ\text{C}$	106			dB
		$R_L = 10\text{ k}\Omega$ , $V_O = \pm 4.7\text{ V}$	98	100		dB
		$-40^\circ\text{C} \leq T_A \leq +125^\circ\text{C}$	90			dB
<b>OUTPUT CHARACTERISTICS</b>						
Output Voltage High	$V_{OH}$	$R_L = 100\text{ k}\Omega$ to GND	4.980	4.990		V
		$-40^\circ\text{C} \leq T_A \leq +125^\circ\text{C}$	4.980			V
		$R_L = 10\text{ k}\Omega$ to GND	4.950	4.960		V
		$-40^\circ\text{C} \leq T_A \leq +125^\circ\text{C}$	4.900			V
Output Voltage Low	$V_{OL}$	$R_L = 100\text{ k}\Omega$ to GND		-4.998	-4.990	V
		$-40^\circ\text{C} \leq T_A \leq +125^\circ\text{C}$			-4.980	V
		$R_L = 10\text{ k}\Omega$ to GND		-4.990	-4.980	V
		$-40^\circ\text{C} \leq T_A \leq +125^\circ\text{C}$			-4.975	V
Short-Circuit Limit	$I_{SC}$	Source/sink		$\pm 20$		mA
Closed-Loop Impedance	$Z_{OUT}$	$f = 1\text{ MHz}$ , $A_V = 1$		77		$\Omega$
<b>POWER SUPPLY</b>						
Power Supply Rejection Ratio	PSRR	$V_{SY} = 2.7\text{ V to }36\text{ V}$	108	126		dB
		$-40^\circ\text{C} \leq T_A \leq +125^\circ\text{C}$	100			dB
Supply Current per Amplifier	$I_{SY}$	$I_O = 0\text{ mA}$		180	225	$\mu\text{A}$
		$-40^\circ\text{C} \leq T_A \leq +125^\circ\text{C}$			300	$\mu\text{A}$
<b>DYNAMIC PERFORMANCE</b>						
Slew Rate	SR	$R_L = 100\text{ k}\Omega$ , $C_L = 30\text{ pF}$		0.46		V/ $\mu\text{s}$
Settling Time	$t_s$	To 0.01%		22		$\mu\text{s}$
Gain Bandwidth Product	GBP			1.22		MHz
Phase Margin	$\Phi_M$			70		Degrees
<b>NOISE PERFORMANCE</b>						
Voltage Noise	$e_n$ p-p	0.1 Hz to 10 Hz		0.8		$\mu\text{V p-p}$
Voltage Noise Density	$e_n$	$f = 1\text{ kHz}$		24		nV/ $\sqrt{\text{Hz}}$

$V_{SY} = \pm 15.0\text{ V}$ ,  $V_{CM} = 0.0\text{ V}$ ,  $V_O = 0.0\text{ V}$ ,  $T_A = 25^\circ\text{C}$ , unless otherwise noted.

Table 4.

Parameter	Symbol	Test Conditions/Comments	Min	Typ	Max	Unit
INPUT CHARACTERISTICS						
Offset Voltage	$V_{OS}$		-250	-35	+250	$\mu\text{V}$
		ADA4091-4 LFCSP package	-400	-40	+400	$\mu\text{V}$
		$-40^\circ\text{C} \leq T_A \leq +125^\circ\text{C}$	-600		+600	$\mu\text{V}$
Offset Voltage Drift	$\Delta V_{OS}/\Delta T$			3.0		$\mu\text{V}/^\circ\text{C}$
Input Bias Current	$I_b$		-60	-50		nA
		$-40^\circ\text{C} \leq T_A \leq +85^\circ\text{C}$	-80		+80	nA
Input Offset Current	$I_{OS}$	$-40^\circ\text{C} \leq T_A \leq +125^\circ\text{C}$	-510		+510	nA
		$-40^\circ\text{C} \leq T_A \leq +85^\circ\text{C}$	-3	0.5	+3	nA
		$-40^\circ\text{C} \leq T_A \leq +125^\circ\text{C}$	-10		+10	nA
Input Voltage Range			-140		+140	nA
Common-Mode Rejection Ratio	CMRR	$V_{CM} = -14.85\text{ V to }+14.85\text{ V}$	104	121		dB
		$-40^\circ\text{C} \leq T_A \leq +125^\circ\text{C}$	95			dB
Large Signal Voltage Gain	$A_{VO}$	$R_L = 100\text{ k}\Omega$ , $V_O = \pm 14.7\text{ V}$	116	119		dB
		$-40^\circ\text{C} \leq T_A \leq +125^\circ\text{C}$	108			dB
		$R_L = 10\text{ k}\Omega$ , $V_O = \pm 14.7\text{ V}$	102	104		dB
		$-40^\circ\text{C} \leq T_A \leq +125^\circ\text{C}$	93			dB
OUTPUT CHARACTERISTICS						
Output Voltage High	$V_{OH}$	$R_L = 100\text{ k}\Omega$ to GND	14.975	14.980		V
		$-40^\circ\text{C} \leq T_A \leq +125^\circ\text{C}$	14.950			V
		$R_L = 10\text{ k}\Omega$ to GND	14.900	14.920		V
		$-40^\circ\text{C} \leq T_A \leq +125^\circ\text{C}$	14.800			V
Output Voltage Low	$V_{OL}$	$R_L = 100\text{ k}\Omega$ to GND		-14.996	-14.990	V
		$-40^\circ\text{C} \leq T_A \leq +125^\circ\text{C}$			-14.985	V
		$R_L = 10\text{ k}\Omega$ to GND		-14.975	-14.950	V
		$-40^\circ\text{C} \leq T_A \leq +125^\circ\text{C}$			-14.940	V
Short-Circuit Limit	$I_{SC}$	Source/sink		$\pm 20$		mA
Closed-Loop Impedance	$Z_{OUT}$	$f = 1\text{ MHz}$ , $A_V = 1$		71		$\Omega$
POWER SUPPLY						
Power Supply Rejection Ratio	PSRR	$V_{SY} = 2.7\text{ V to }36\text{ V}$	108	126		dB
		$-40^\circ\text{C} \leq T_A \leq +125^\circ\text{C}$	100			dB
Supply Current per Amplifier	$I_{SY}$	$I_O = 0\text{ mA}$		200	250	$\mu\text{A}$
		$-40^\circ\text{C} \leq T_A \leq +125^\circ\text{C}$			350	$\mu\text{A}$
DYNAMIC PERFORMANCE						
Slew Rate	SR	$R_L = 100\text{ k}\Omega$ , $C_L = 30\text{ pF}$		0.46		V/ $\mu\text{s}$
Settling Time	$t_s$	To 0.01%		22		$\mu\text{s}$
Gain Bandwidth Product	GBP			1.27		MHz
Phase Margin	$\Phi_M$			72		Degrees
Channel Separation	CS	$f = 1\text{ kHz}$		100		dB
NOISE PERFORMANCE						
Voltage Noise	$e_n$ p-p	0.1 Hz to 10 Hz		0.8		$\mu\text{V p-p}$
Voltage Noise Density	$e_n$	$f = 1\text{ kHz}$		25		nV/ $\sqrt{\text{Hz}}$

## ABSOLUTE MAXIMUM RATINGS

Table 5.

Parameter	Rating
Supply Voltage	36 V
Input Voltage	Refer to the Input Overvoltage Protection section
Differential Input Voltage <sup>1</sup>	$\pm V_{SY}$
Input Current	$\pm 5$ mA
Output Short-Circuit Duration to GND	Indefinite
Storage Temperature Range	-65°C to +150°C
Operating Temperature Range	-40°C to +125°C
Junction Temperature Range	-65°C to +150°C
Lead Temperature (Soldering, 60 sec)	300°C

<sup>1</sup> Input current must be limited to  $\pm 5$  mA.

Stresses at or above those listed under Absolute Maximum Ratings may cause permanent damage to the product. This is a stress rating only; functional operation of the product at these or any other conditions above those indicated in the operational section of this specification is not implied. Operation beyond the maximum operating conditions for extended periods may affect product reliability.

## THERMAL RESISTANCE

$\theta_{JA}$  is specified for the device soldered on a 4-layer JEDEC standard PCB with zero airflow. The exposed pad is soldered to the application board.

Table 6. Thermal Resistance

Package Type	$\theta_{JA}$	$\theta_{JC}$	Unit
8-Lead SOIC (R-8)	155	45	°C/W
14-Lead TSSOP (RU-14)	112	35	°C/W
8-Lead LFCSP (CP-8-21)	75	12	°C/W
16-Lead LFCSP (CP-16-17)	55	14	°C/W

## ESD CAUTION



**ESD (electrostatic discharge) sensitive device.** Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

TYPICAL PERFORMANCE CHARACTERISTICS

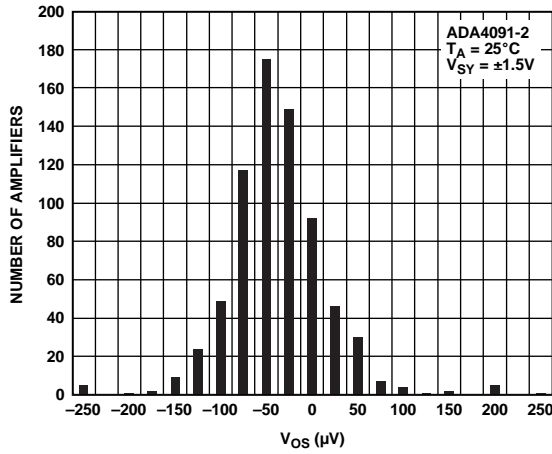


Figure 5. Input Offset Voltage Distribution

07671-034

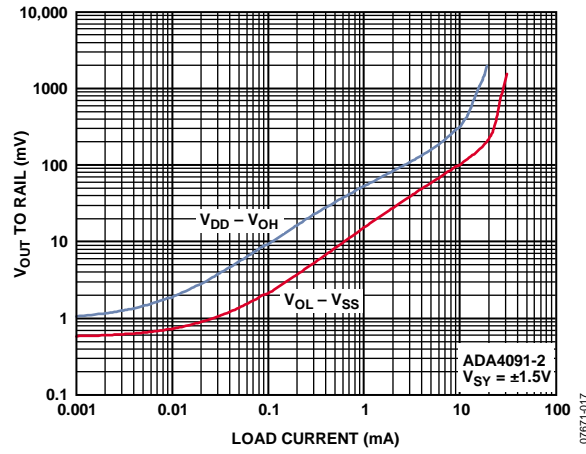


Figure 8. Dropout Voltage vs. Load Current

07671-017

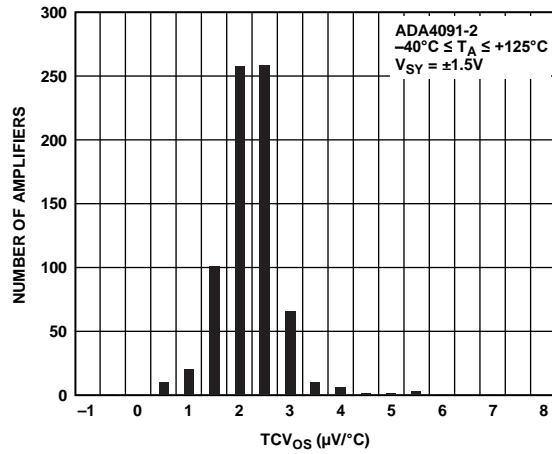


Figure 6. TCV<sub>0s</sub> Distribution

07671-035

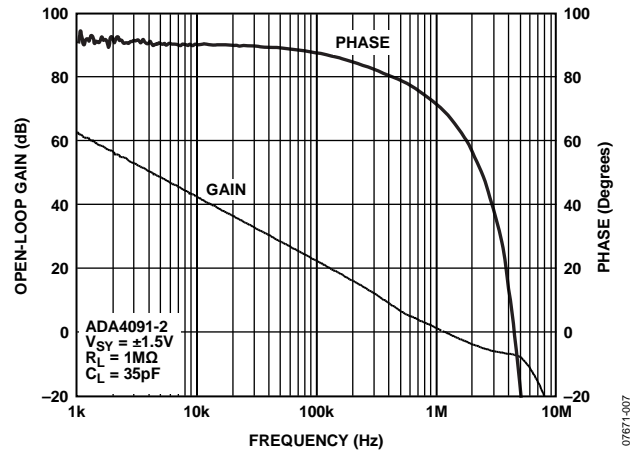


Figure 9. Open-Loop Gain and Phase vs. Frequency

07671-007

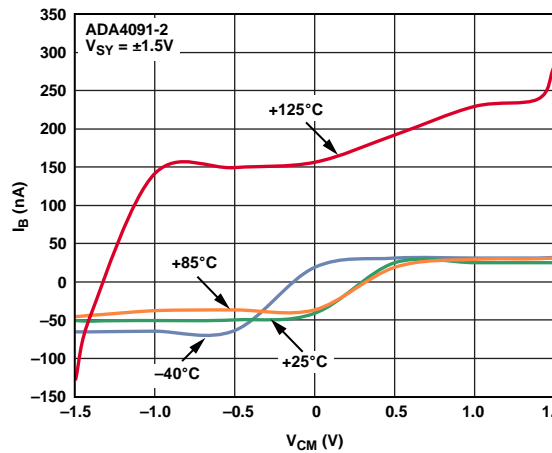


Figure 7. Input Bias Current vs. Common-Mode Voltage

07671-033

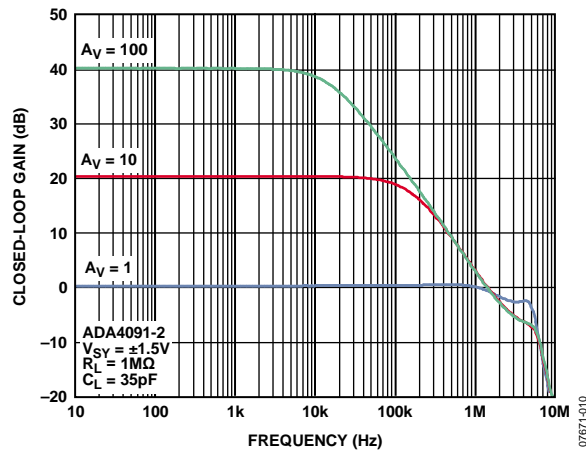


Figure 10. Closed-Loop Gain vs. Frequency

07671-010

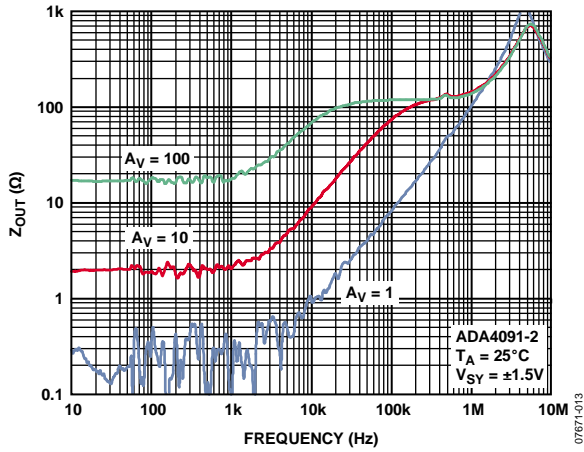


Figure 11. Output Impedance vs. Frequency

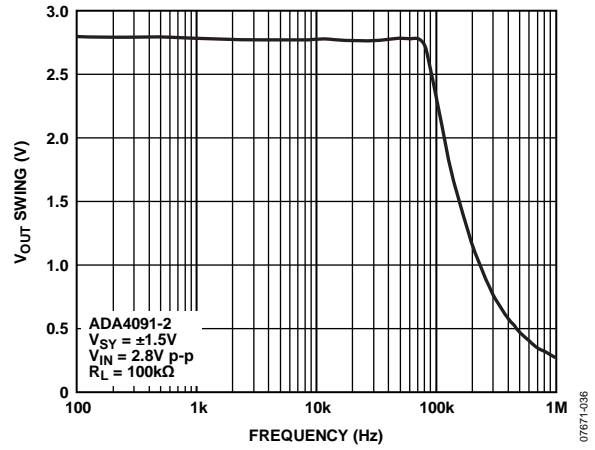


Figure 14. Output Swing vs. Frequency

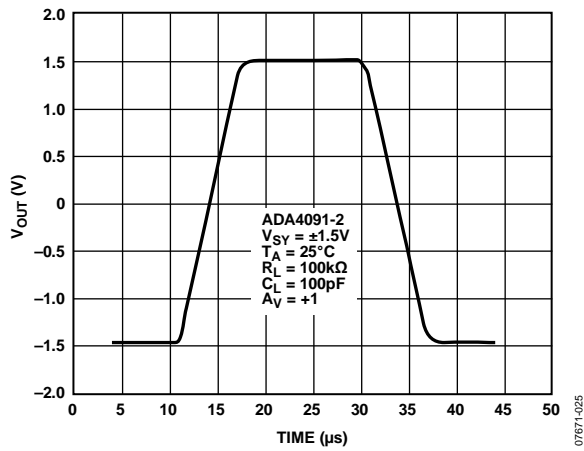


Figure 12. Large Signal Transient Response

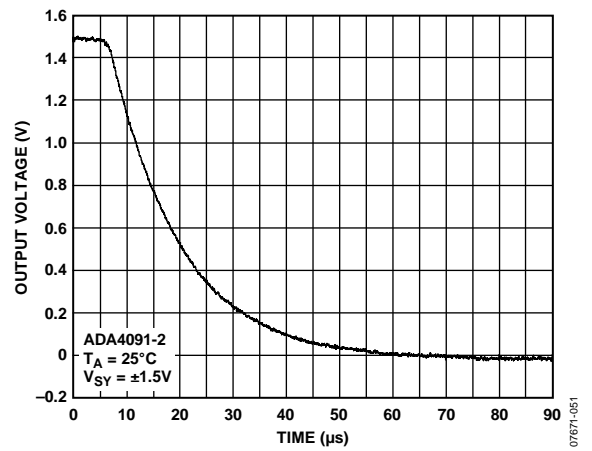


Figure 15. Positive Overload Recovery

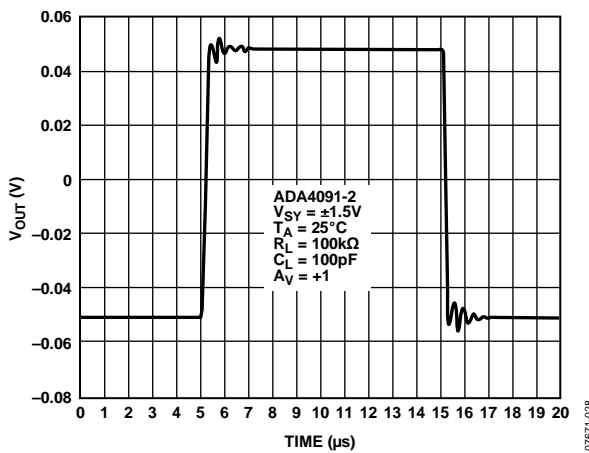


Figure 13. Small Signal Transient Response

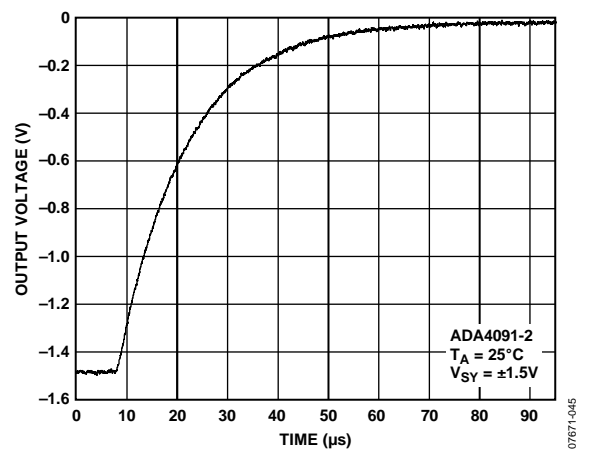


Figure 16. Negative Overload Recovery



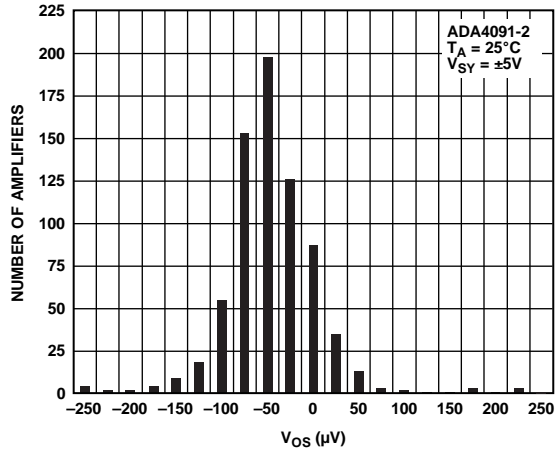


Figure 17. Input Offset Voltage Distribution

07671-037

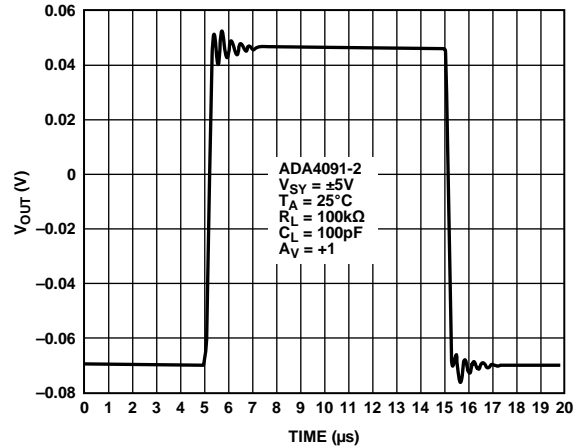


Figure 20. Small Signal Transient Response

07671-039

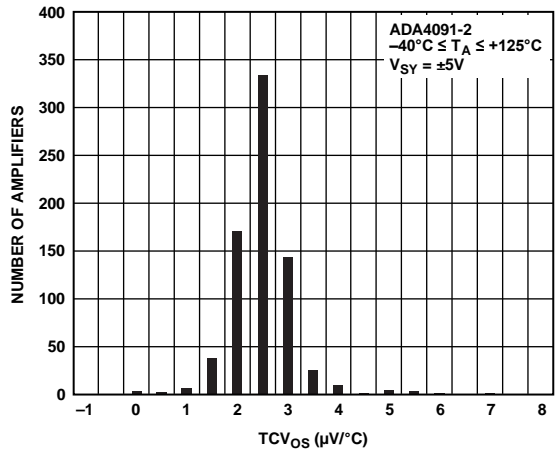


Figure 18. TCV<sub>OS</sub> Distribution

07671-038

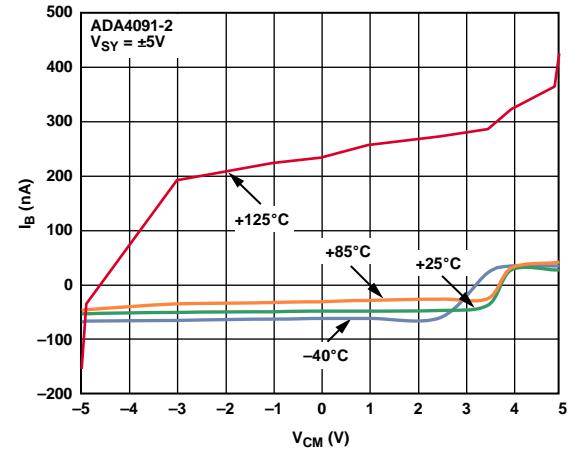


Figure 21. Input Bias Current vs. Common-Mode Voltage

07671-032

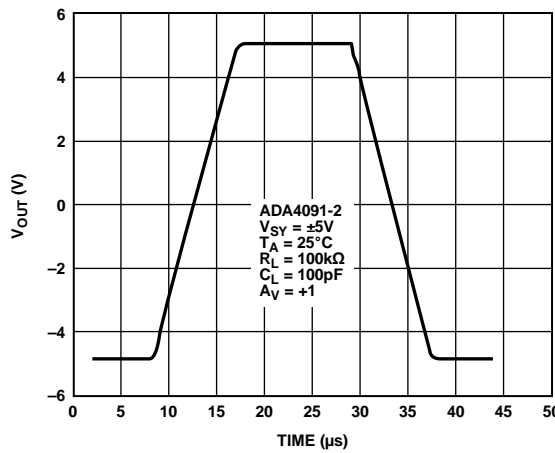


Figure 19. Large Signal Transient Response

07671-026

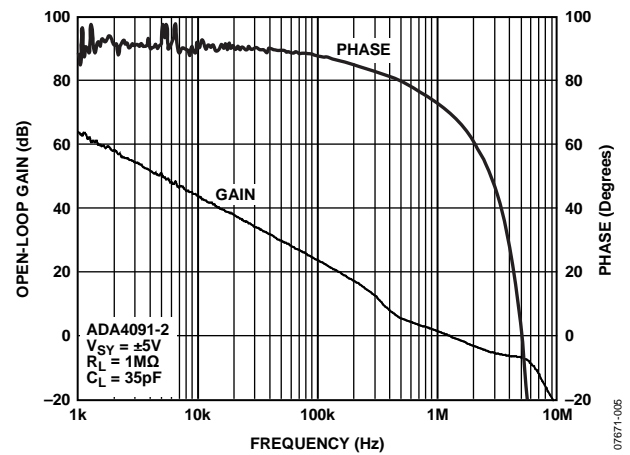


Figure 22. Open-Loop Gain and Phase vs. Frequency

07671-095

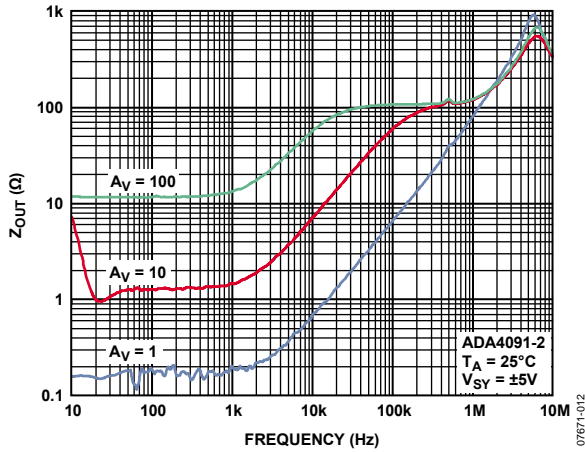


Figure 23. Output Impedance vs. Frequency

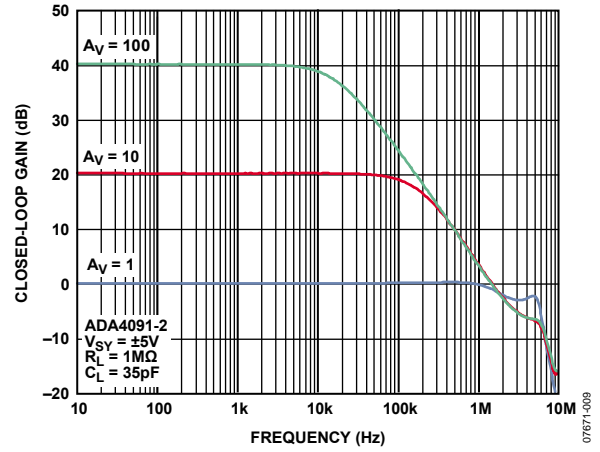


Figure 26. Closed-Loop Gain vs. Frequency

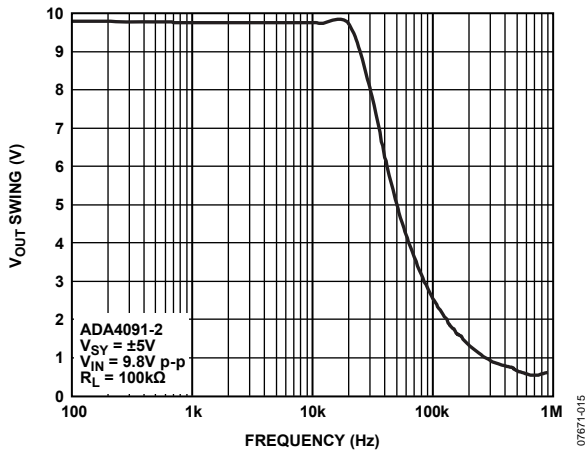


Figure 24. Output Voltage Swing vs. Frequency

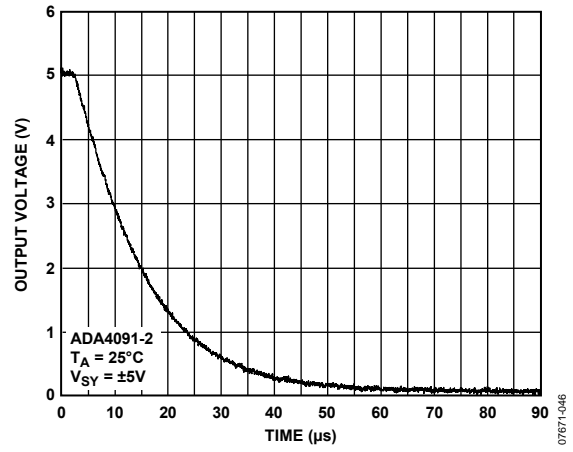


Figure 27. Positive Overload Recovery

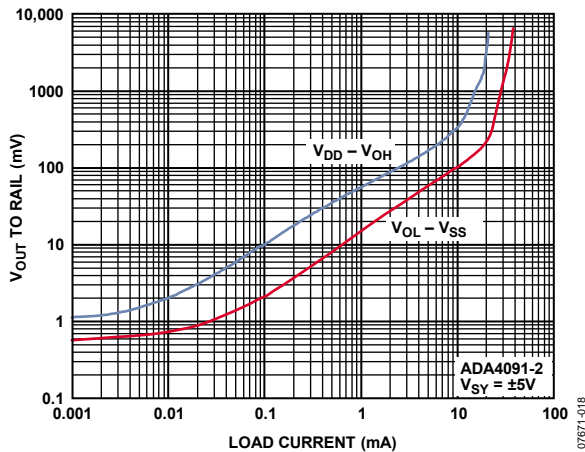


Figure 25. Dropout Voltage vs. Load Current

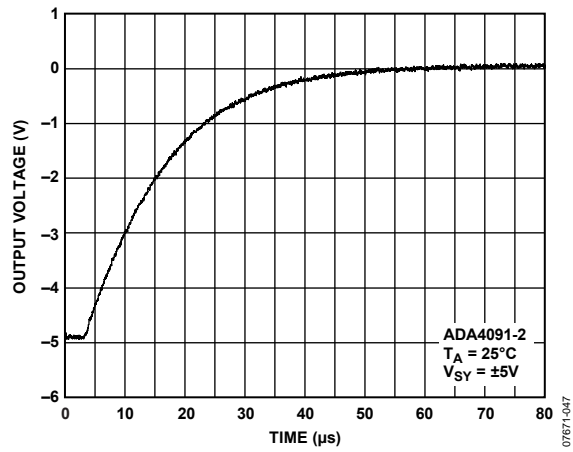


Figure 28. Negative Overload Recovery

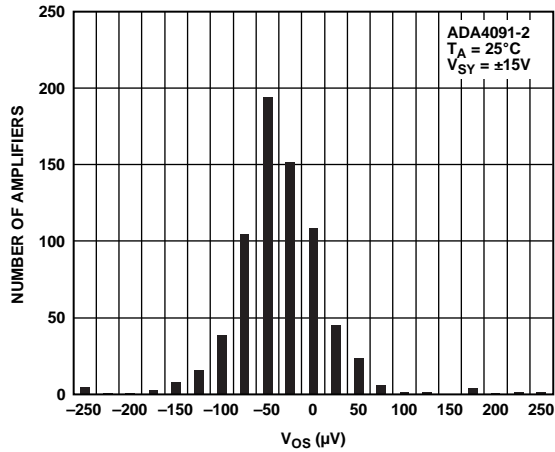


Figure 29. Input Offset Voltage Distribution

07671-041

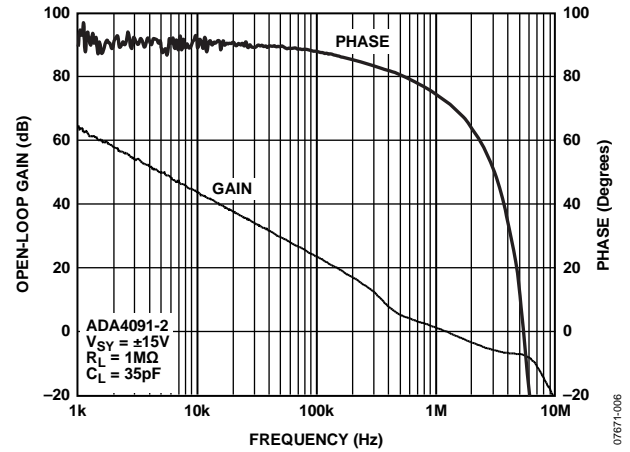


Figure 32. Open-Loop Gain and Phase vs. Frequency

07671-006

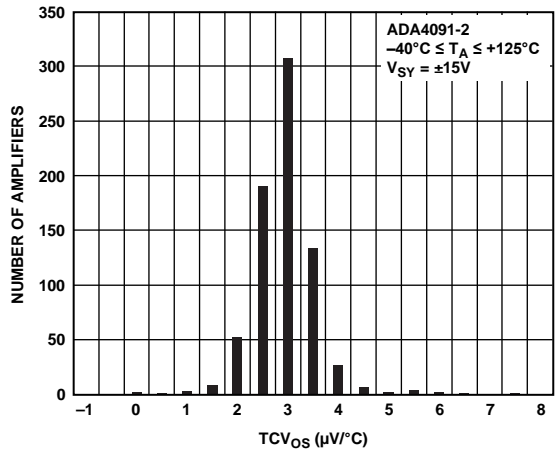


Figure 30.  $TCV_{OS}$  Distribution

07671-042

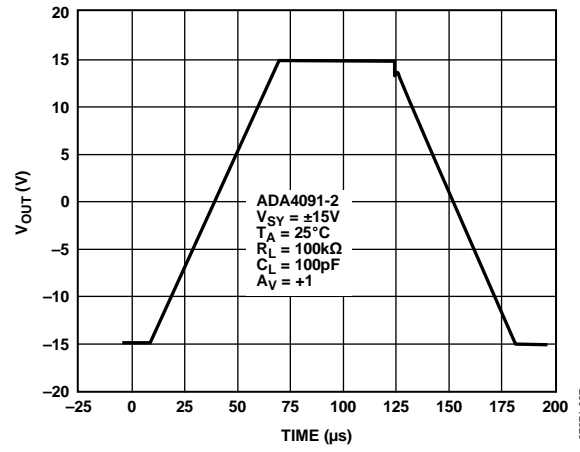


Figure 33. Large Signal Transient Response

07671-027

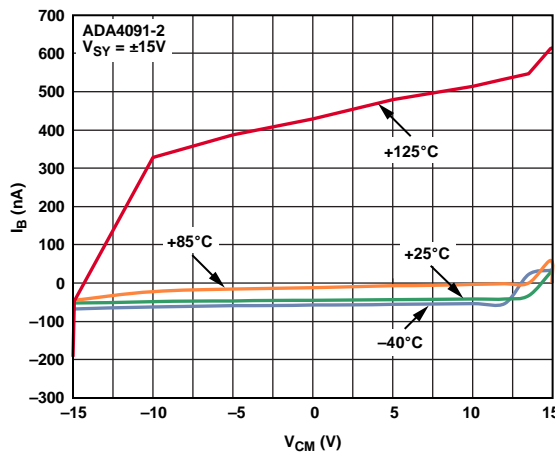


Figure 31. Input Bias Current vs. Common-Mode Voltage

07671-031

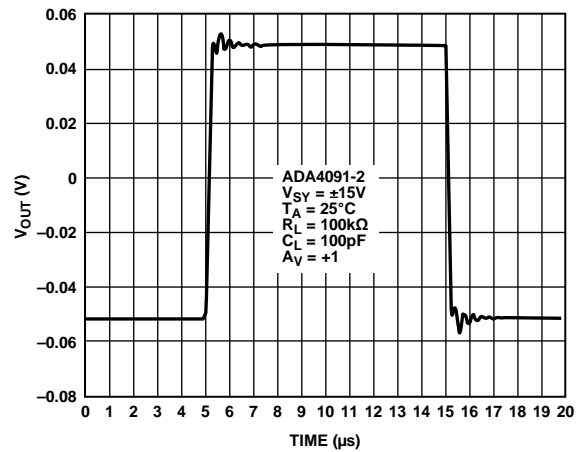


Figure 34. Small Signal Transient Response

07671-030

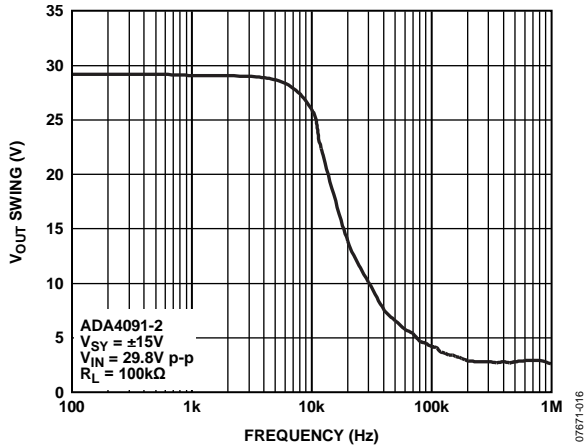


Figure 35. Output Voltage Swing vs. Frequency

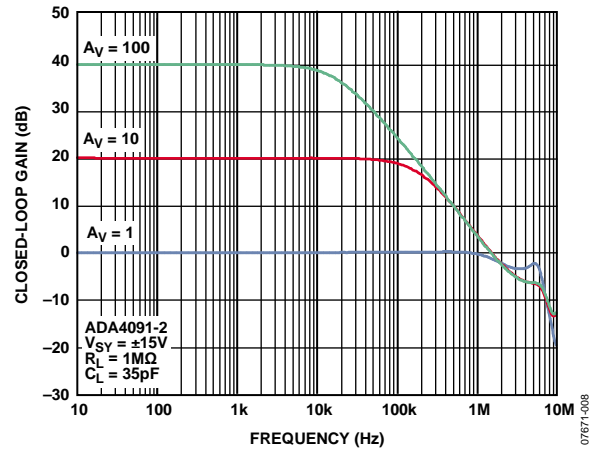


Figure 38. Closed-Loop Gain vs. Frequency

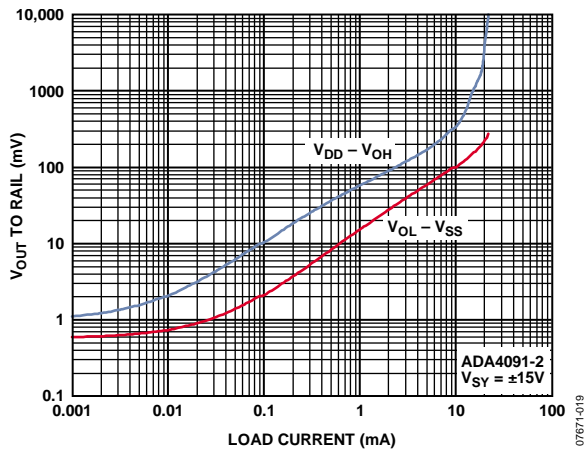


Figure 36. Dropout Voltage vs. Load Current

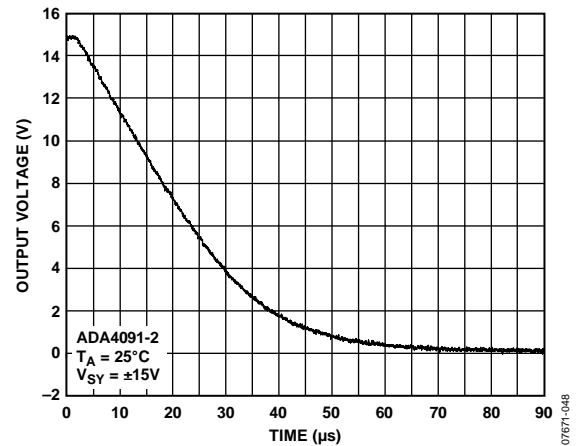


Figure 39. Positive Overload Recovery

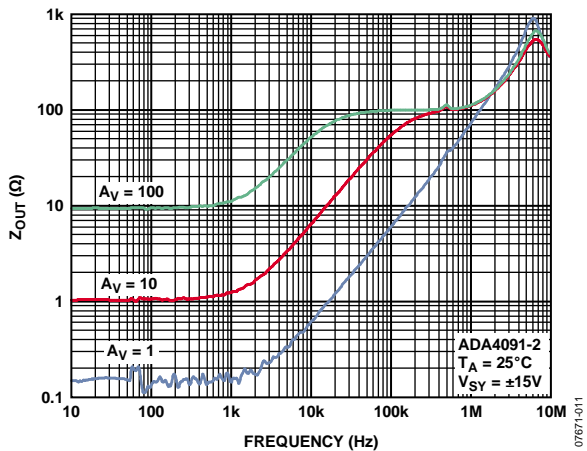


Figure 37. Output Impedance vs. Frequency

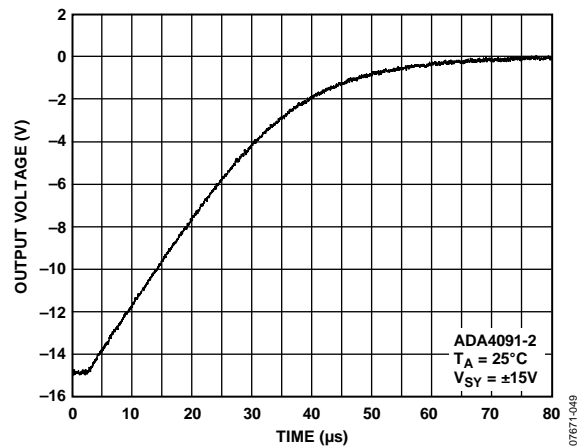


Figure 40. Negative Overload Recovery

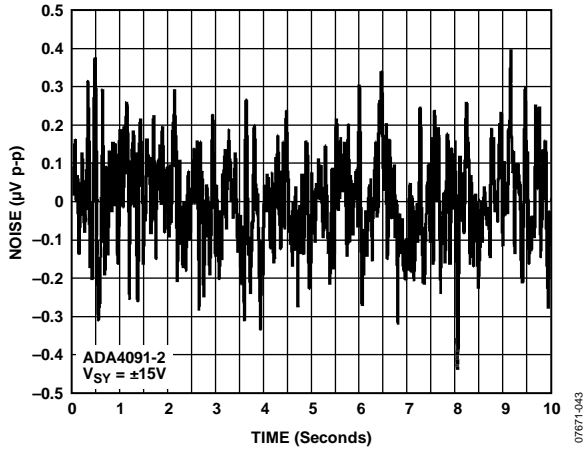


Figure 41. Peak-to-Peak Voltage Noise

07671-043

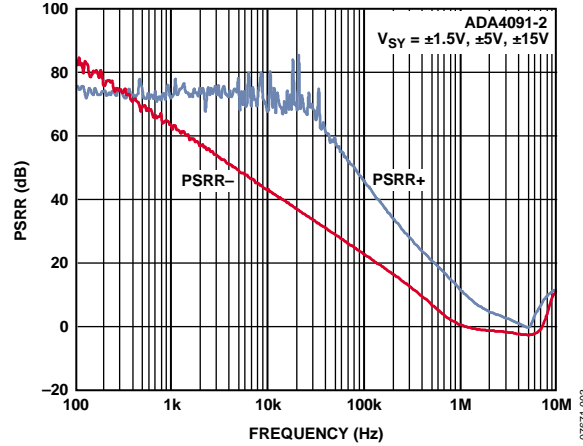


Figure 44. PSRR vs. Frequency

07671-003

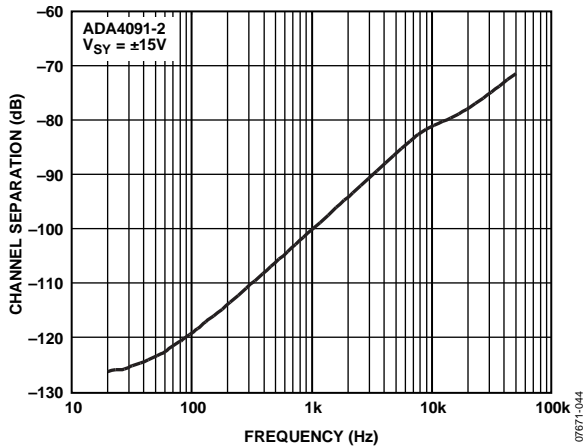


Figure 42. Channel Separation vs. Frequency

07671-044

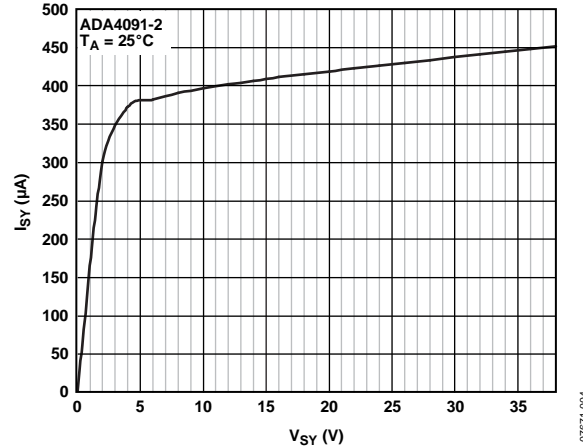


Figure 45. Supply Current vs. Supply Voltage

07671-004

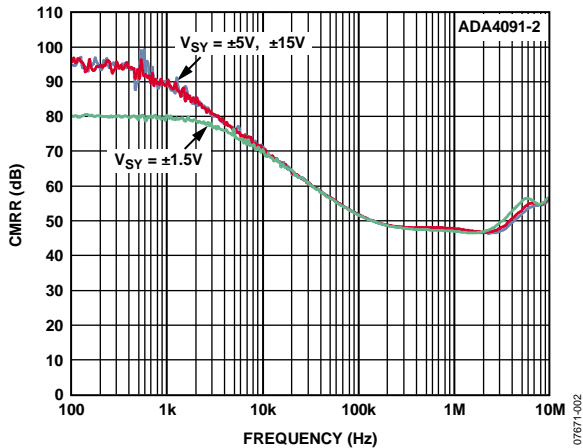


Figure 43. CMRR vs. Frequency

07671-002

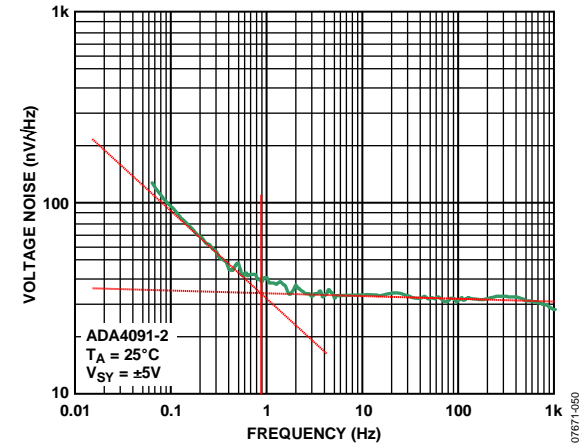


Figure 46. Voltage Noise Density

07671-050

## THEORY OF OPERATION

The ADA4091-2/ADA4091-4 is a single-supply, micropower amplifier featuring rail-to-rail inputs and outputs. To achieve wide input and output ranges, these amplifiers employ unique input and output stages.

### INPUT STAGE

In Figure 47, the input stage comprises two differential pairs, a PNP pair (PNP input stage) and an NPN pair (NPN input stage). These input stages do not work in parallel. Instead, only one stage is on for any given input common-mode signal level. The PNP stage (Transistor Q1 and Transistor Q2) is required to ensure that the amplifier remains in the linear region when the input voltage approaches and reaches the negative rail. Alternatively, the NPN stage (Transistor Q5 and Transistor Q6) is needed for input voltages up to, and including, the positive rail.

For the majority of the input common-mode range, the PNP stage is active, as shown in Figure 7, Figure 21, and Figure 31. Notice that the bias current switches direction at approximately 1.5 V below the positive rail. At voltages below this level, the bias current flows out of the ADA4091-2/ADA4091-4 input, from the PNP input stage. Above this voltage, however, the bias current enters the device, due to the NPN stage. The actual mechanism within the amplifier for switching between the input stages comprises Transistor Q3, Transistor Q4, and Transistor Q7. As the input common-mode voltage increases, the emitters of Q1 and Q2 follow that voltage plus a diode drop.

Eventually, the emitters of Q1 and Q2 are high enough to turn on Q3, which diverts the tail current away from the PNP input stage, turning it off. The tail current of the PNP pair is diverted to the Q4/Q7 current mirror to activate the NPN input stage.

A common practice in bipolar amplifiers to protect the input transistors from large differential voltages is to include series resistors and differential diodes. See Figure 48 for the full input protection circuitry. These diodes turn on whenever the differential voltage exceeds approximately 0.6 V. In this condition, current flows between the input pins, limited only by the two 5 kΩ resistors. Evaluate each application carefully to make sure that the increase in current does not affect performance.

### OUTPUT STAGE

The output stage in the ADA4091-2/ADA4091-4 device uses a PNP and an NPN transistor, as do most output stages. However, Q32 and Q33, the output transistors, connect with their collectors to the output pin to achieve the rail-to-rail output swing.

As the output voltage approaches either the positive or negative rail, these transistors begin to saturate. Thus, the final limit on output voltage is the saturation voltage of these transistors, which is about 50 mV. The output stage has inherent gain arising from the transistor output impedance, as well as any external load impedance; consequently, the open-loop gain of the operational amplifier is dependent on the load resistance and decreases when the output voltage is close to either rail.

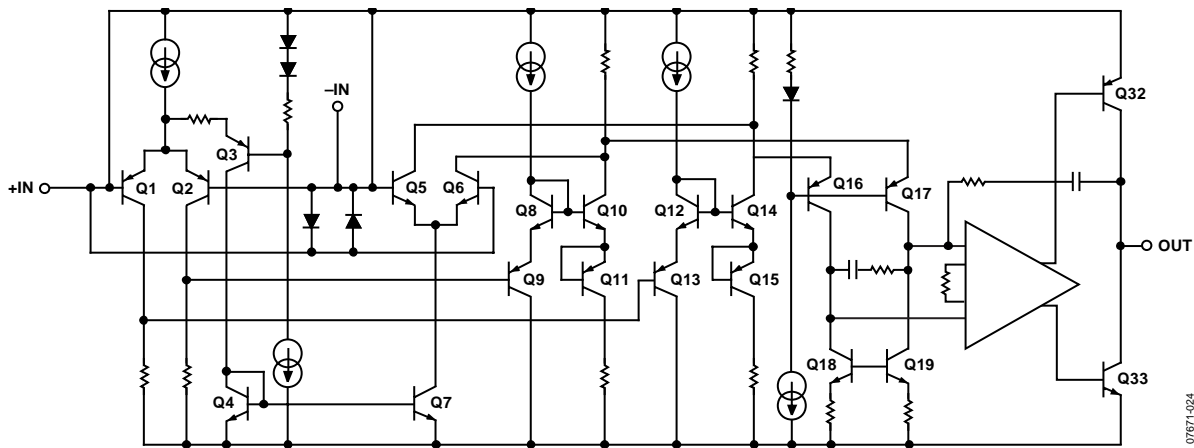


Figure 47. Simplified Schematic Without Input Protection (see Figure 48)

07671-024

**INPUT OVERVOLTAGE PROTECTION**

The ADA4091-2/ADA4091-4 has two different ESD circuits for enhanced protection, as shown in Figure 48.

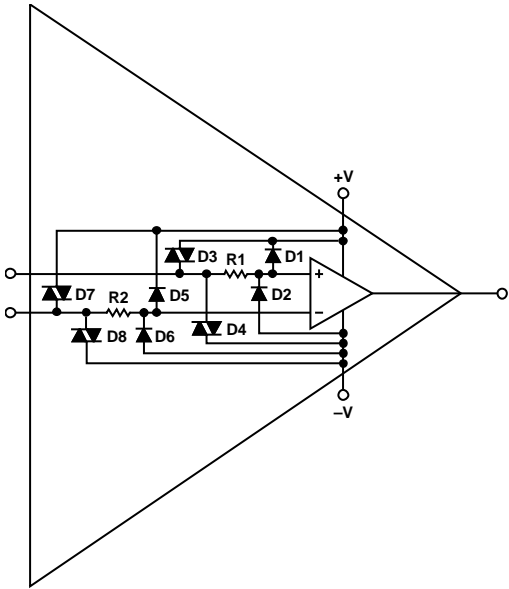


Figure 48. Complete Input Protection Network

One circuit is a series resistor of 5 kΩ to the internal inputs and diodes (D1 and D2 or D5 and D6) from the internal inputs to the supply rails. The other protection circuit is a circuit with two DIACs (D3 and D4 or D7 and D8) to the supply rails. A DIAC can be considered a bidirectional Zener diode with a transfer characteristic, as shown in Figure 49.

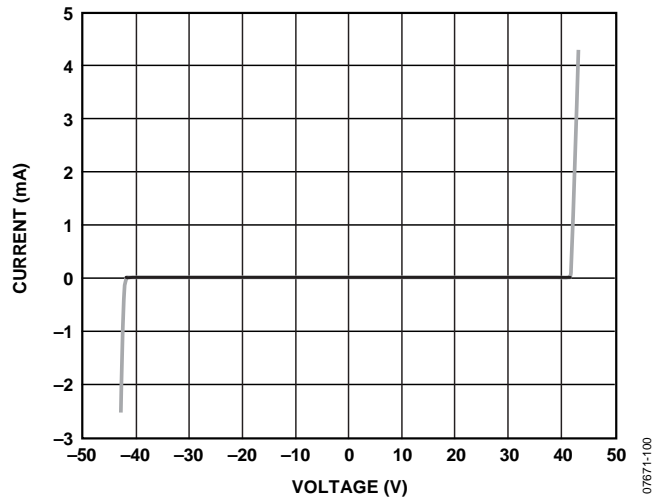


Figure 49. DIAC Transfer Characteristic

For a worst-case design analysis, consider two cases. The ADA4091-2/ADA4091-4 has a normal ESD structure from the internal operational amplifier inputs to the supply rails. In addition, it has 42 V DIACs from the external inputs to the rails, as shown in Figure 47.

Therefore, two conditions need to be considered to determine which case is the limiting factor.

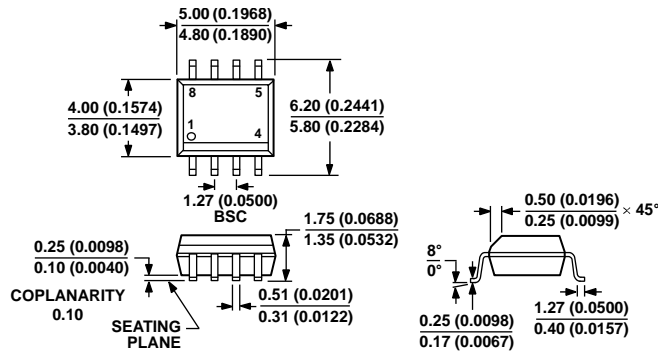
- Condition 1. Consider, for example, that when operating on ±15 V, the inputs can go +42 V above the negative supply rail. With the -V pin equal to -15 V, +42 V above this supply (the negative supply) is +27 V.
- Condition 2. There is a restriction on the input current of 5 mA through a 5 kΩ resistor to the ESD structure to the positive rail. In Condition 1, +27 V through the 5 kΩ resistor to +15 V gives a current of 2.4 mA. Thus, the DIAC is the limiting factor. If the ADA4091-2/ADA4091-4 supply voltages are changed to ±5 V, then -5 V + 42 V = +37 V. However, +5 V + (5 kΩ × 5 mA) = 30 V. Thus, the normal resistor diode structure is the limitation when running on lower supply voltages.

Additional resistance can be added externally in series with each input to protect against higher peak voltages; however, the additional thermal noise of the resistors must be considered.

The flatband voltage noise of the ADA4091-2/ADA4091-4 is approximately 24 nV/√Hz, and a 5 kΩ resistor has a noise of 9 nV/√Hz. Adding an additional 5 kΩ resistor increases the total noise by less than 15% root sum square (rss). Therefore, maintain resistor values below this value (5 kΩ) when overall noise performance is critical.

Note that this represents input protection under abnormal conditions only. The correct amplifier operation input voltage range (IVR) is specified in Table 2, Table 3, and Table 4.

OUTLINE DIMENSIONS

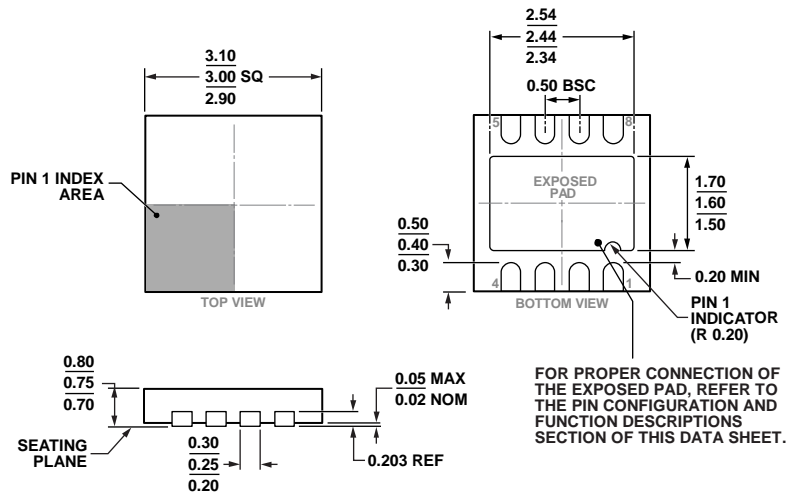


COMPLIANT TO JEDEC STANDARDS MS-012-AA  
 CONTROLLING DIMENSIONS ARE IN MILLIMETERS; INCH DIMENSIONS  
 (IN PARENTHESES) ARE ROUNDED-OFF MILLIMETER EQUIVALENTS FOR  
 REFERENCE ONLY AND ARE NOT APPROPRIATE FOR USE IN DESIGN.

012407-A

Figure 50. 8-Lead Standard Small Outline Package [SOIC\_N]  
 Narrow Body  
 (R-8)

Dimensions shown in millimeters and (inches)

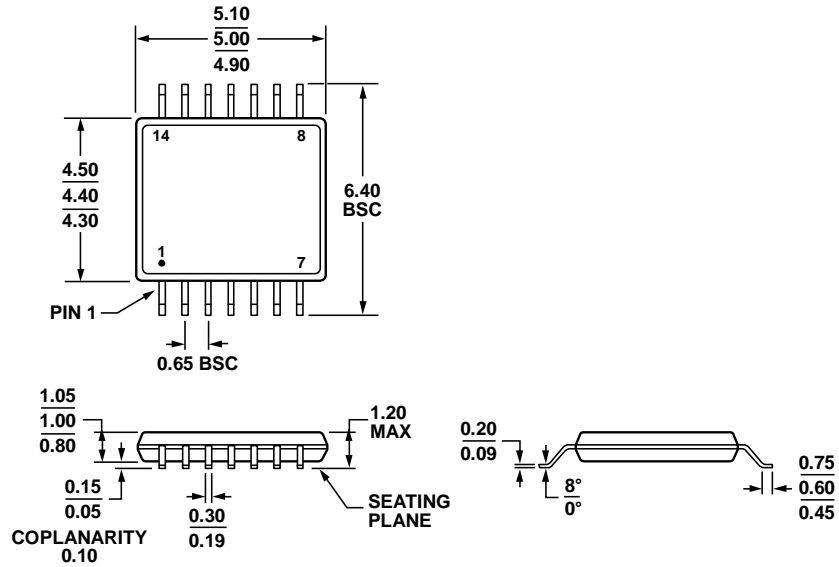


FOR PROPER CONNECTION OF  
 THE EXPOSED PAD, REFER TO  
 THE PIN CONFIGURATION AND  
 FUNCTION DESCRIPTIONS  
 SECTION OF THIS DATA SHEET.

Figure 51. 8-Lead Lead Frame Chip Scale Package [LFCSP]  
 3 mm x 3 mm Body and 0.75 mm Package Height  
 (CP-8-21)

Dimensions shown in millimeters



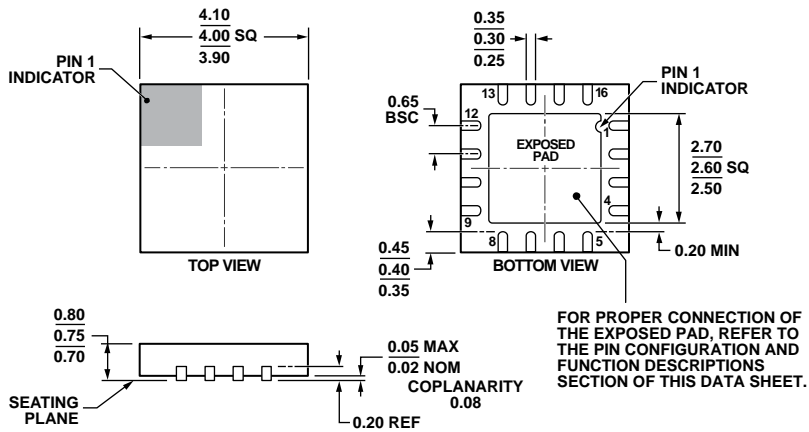


COMPLIANT TO JEDEC STANDARDS MO-153-AB-1

Figure 52. 14-Lead Thin Shrink Small Outline Package [TSSOP]  
Narrow Body  
(RU-14)

Dimensions shown in millimeters

061908-A



COMPLIANT TO JEDEC STANDARDS MO-220-WGGC.

Figure 53. 16-Lead Lead Frame Chip Scale Package [LFCSF]  
4 mm x 4 mm Body and 0.75 mm Package Height  
(CP-16-17)

Dimensions are millimeters

0816-2010-C

**ORDERING GUIDE**

<b>Model<sup>1</sup></b>	<b>Temperature Range</b>	<b>Package Description</b>	<b>Package Option</b>	<b>Branding</b>
ADA4091-2ARZ	-40°C to +125°C	8-Lead Standard Small Outline Package [SOIC_N]	R-8	
ADA4091-2ARZ-R7	-40°C to +125°C	8-Lead Standard Small Outline Package [SOIC_N]	R-8	
ADA4091-2ARZ-RL	-40°C to +125°C	8-Lead Standard Small Outline Package [SOIC_N]	R-8	
ADA4091-2ACPZ-R2	-40°C to +125°C	8-Lead Lead Frame Chip Scale Package [LFCSP]	CP-8-21	A1Z
ADA4091-2ACPZ-R7	-40°C to +125°C	8-Lead Lead Frame Chip Scale Package [LFCSP]	CP-8-21	A1Z
ADA4091-2ACPZ-RL	-40°C to +125°C	8-Lead Lead Frame Chip Scale Package [LFCSP]	CP-8-21	A1Z
ADA4091-4ARUZ	-40°C to +125°C	14-Lead Thin Shrink Small Outline Package [TSSOP]	RU-14	
ADA4091-4ARUZ-RL	-40°C to +125°C	14-Lead Thin Shrink Small Outline Package [TSSOP]	RU-14	
ADA4091-4ACPZ-R2	-40°C to +125°C	16-Lead Lead Frame Chip Scale Package [LFCSP]	CP-16-17	
ADA4091-4ACPZ-R7	-40°C to +125°C	16-Lead Lead Frame Chip Scale Package [LFCSP]	CP-16-17	
ADA4091-4ACPZ-RL	-40°C to +125°C	16-Lead Lead Frame Chip Scale Package [LFCSP]	CP-16-17	

<sup>1</sup>Z = RoHS Compliant Part.

**NOTES**

NOTES